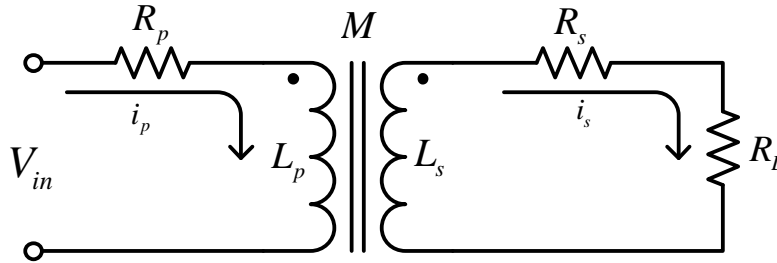


Simple Transformer Analysis
7-28-08



KVL on the primary side gives:

$$V_{in} = R_p i_p + \varepsilon_p \quad (1.1)$$

where the primary winding EMF is given by (see appendix A):

$$\varepsilon_p = L_p \frac{di_p}{dt} - M \frac{di_s}{dt} \quad (1.2)$$

KVL on the secondary side gives:

$$\varepsilon_s = (R_s + R_L) i_s \quad (1.3)$$

where the secondary EMF is given by:

$$\varepsilon_s = M \frac{di_p}{dt} - L_s \frac{di_s}{dt} \quad (1.4)$$

Combining (1.1) and (1.2):

$$V_{in} = R_p i_p + L_p \frac{di_p}{dt} - M \frac{di_s}{dt} \quad (1.5)$$

Combining (1.3) and (1.4):

$$M \frac{di_p}{dt} - L_s \frac{di_s}{dt} = (R_s + R_L) i_s \quad (1.6)$$

Note that the mutual inductance is given by:

$$M = \sqrt{L_p L_s} \quad (1.7)$$

Rewriting the winding EMF's:

$$\varepsilon_p = L_p \frac{di_p}{dt} - \sqrt{L_p L_s} \frac{di_s}{dt} \quad (1.8)$$

$$\varepsilon_s = \sqrt{L_p L_s} \frac{di_p}{dt} - L_s \frac{di_s}{dt} \quad (1.9)$$

Note that:

$$\sqrt{\frac{L_s}{L_p}} \cdot \varepsilon_p = \sqrt{L_p L_s} \frac{di_p}{dt} - L_s \frac{di_s}{dt} \Rightarrow \varepsilon_s = \sqrt{\frac{L_s}{L_p}} \cdot \varepsilon_p \quad (1.10)$$

In terms of number of turns in each of the windings, this becomes:

$$\boxed{\varepsilon_s = \frac{N_s}{N_p} \varepsilon_p} \quad (1.11)$$

Plugging (1.11) into (1.1), we get:

$$V_{in} = R_p i_p + \frac{N_p}{N_s} \varepsilon_s \quad (1.12)$$

And plugging (1.3) into this, we get:

$$V_{in} = R_p i_p + \frac{N_p}{N_s} (R_s + R_L) i_s \quad (1.13)$$

Taking the Laplace transform of (1.5) and (1.6), we get:

$$V_{in} = (L_p s + R_p) i_p - (M s) i_s \quad (1.14)$$

and

$$(M s) i_p = [L_s s + (R_s + R_L)] i_s \quad (1.15)$$

Plugging (1.15) into (1.14) to eliminate i_s :

$$\begin{aligned} V_{in} &= \left[(L_p s + R_p) - \frac{M^2 s^2}{L_s s + (R_s + R_L)} \right] i_p \\ &= \frac{(L_p s + R_p) [L_s s + (R_s + R_L)] - M^2 s^2}{L_s s + (R_s + R_L)} i_p \\ &= \frac{(L_p L_s - M^2) s^2 + [L_p (R_s + R_L) + L_s R_p] s + (R_s + R_L) R_p}{L_s s + (R_s + R_L)} i_p \\ &= \frac{[L_p (R_s + R_L) + L_s R_p] s + (R_s + R_L) R_p}{L_s s + (R_s + R_L)} i_p \end{aligned} \quad (1.16)$$

The effective impedance of the circuit as seen by the input voltage source is:

$$\boxed{Z_{eff}(s) = \frac{V_{in}(s)}{i_p(s)} = \frac{[L_p (R_s + R_L) + L_s R_p] s + (R_s + R_L) R_p}{L_s s + (R_s + R_L)}} \quad (1.17)$$

Looking at this TF for extremely low and high frequencies:

$$\lim_{s \rightarrow 0} Z_{eff}(s) = R_p \quad (1.18)$$

and

$$\begin{aligned} \lim_{s \rightarrow \infty} Z_{eff}(s) &= \frac{L_p (R_s + R_L) + L_s R_p}{L_s} \\ &= \frac{L_p}{L_s} (R_s + R_L) + R_p \\ &= \frac{N_p^2}{N_s^2} (R_s + R_L) + R_p \end{aligned} \quad (1.19)$$

Appendix A – Derivation of Self and Mutual Inductances

The total flux in the core is given by:

$$\Phi = \Phi_p + \Phi_s \quad (1.20)$$

where:

$$\Phi_p = \frac{\mu A_c N_p^2}{\ell_m} i_p \quad (1.21)$$

and

$$\Phi_s = -\frac{\mu A_c N_s^2}{\ell_m} i_s \quad (1.22)$$

The winding EMF's are given by:

$$\varepsilon_p = N_p \frac{d\Phi}{dt} = N_p \left(\frac{d\Phi_p}{dt} + \frac{d\Phi_s}{dt} \right) = \underbrace{\frac{\mu A_c N_p^2}{\ell_m}}_{L_p} \frac{di_p}{dt} - \underbrace{\frac{\mu A_c N_p N_s}{\ell_m}}_{M_{ps}} \frac{di_s}{dt} \quad (1.23)$$

$$\varepsilon_s = N_s \frac{d\Phi}{dt} = N_s \left(\frac{d\Phi_p}{dt} + \frac{d\Phi_s}{dt} \right) = \underbrace{\frac{\mu A_c N_p N_s}{\ell_m}}_{M_{sp}} \frac{di_p}{dt} - \underbrace{\frac{\mu A_c N_s^2}{\ell_m}}_{L_s} \frac{di_s}{dt} \quad (1.24)$$

Note that:

$$M_{ps} = M_{sp} = M \quad (1.25)$$

Rewriting the winding EMF's:

$$\varepsilon_p = L_p \frac{di_p}{dt} - M \frac{di_s}{dt} \quad (1.26)$$

$$\varepsilon_s = M \frac{di_p}{dt} - L_s \frac{di_s}{dt} \quad (1.27)$$